ANALOG CIRCUIT DESIGN USING VCII+

A Thesis Submitted in Partial Fulfillment of the Requirements for the Degree of

MASTER OF TECHNOLOGY

in

CONTROL & INSTRUMENTATION

by

Rishabh Verma 23/C&I/05

Under the Supervision of

Dr. Garima

Dr. Bhavnesh Jaint



Department of Electrical Engineering

DELHI TECHNOLOGICAL UNIVERSITY

(Formerly Delhi College of Engineering) Bawana Road, New Delhi-110042

MAY, 2025

ACKNOWLEDGEMENT

I would like to thank **Dr. Rachna Garg**, the Head of Department, Department of Electrical Engineering, for their continuous support and for providing a conducive environment for research. Their leadership and encouragement have been crucial in facilitating the progress of my work.

I take immense pleasure to express my deep and sincere gratitude to my esteemed supervisor, **Dr. Garima**, for her invaluable guidance and spending of her precious hours for my work. Her excellent cooperation and suggestion through stimulating and beneficial discussions provided me with an impetus to work and made the completion of the thesis possible. I also extend my sincere thanks to my co-supervisor **Dr. Bhavnesh Jaint** for offering invaluable assistance and helping me in my research work which boost up my confidence and pave the way to plan for surveying at the initial stages. She provided me constant encouragement and support in all possible ways.

I also appreciate and heartily thanks to all faculty member of **Department of Electrical Engineering** and all my lab mates for their constant support in my research work. I am also thankful to all my classmates and technical staff for supporting and helping me throughout the research work.

Finally, yet importantly, I would like to pay high regards to lord almighty, my Parents, my family members and my friends for their inspiration and motivation throughout my work and lifting me uphill this phase of life.



DELHI TECHNOLOGICAL UNIVERSITY

(Formerly Delhi College of Engineering) Shahbad Daulatpur, Bawana Road, New Delhi-110042

CANDIDATE'S DECLARATION

I, Rishabh Verma (23/C&I/05) student of M.Tech (Control & Instrumentation), hereby certify that the work which is being presented in the thesis entitled "ANALOG CIRCUIT DESIGN USING VCII+" in partial fulfillment of the requirements for the degree of Master of Technology, submitted in Department of Electrical Engineering, Delhi Technological University, New Delhi is an authentic record of my own work and not copied from any source without proper citation which is carried out under the supervision of Dr. Garima and co-supervision of Dr. Bhavnesh Jaint.

The matter presented in the thesis has not been submitted by me for the award of any other degree of this or any other Institute.

Candidate's Signature

This is to certify that the student has incorporated all the corrections suggested by the examiners in the thesis and the statement made by the candidate is correct to the best of our knowledge.

Signature of Supervisors

Signature of External Examiner

Date: May 2025



DELHI TECHNOLOGICAL UNIVERSITY

(Formerly Delhi College of Engineering) Shahbad Daulatpur, Bawana Road, New Delhi-110042

CERTIFICATE BY THE SUPERVISOR

Certified that **Rishabh Verma** (23/C&I/05) has carried out their research work presented in this thesis entitled "ANALOG CIRCUIT DESIGN USING VCII+" for the award of Master of Technology from Department of Electrical Engineering, Delhi Technological University, New Delhi under our supervision. The thesis embodies results of original work, and studies carried out by the students himself and the contents of the thesis do not form the basis for the award of any other degree to the candidate or to anybody else from this or any other University/Institution.

Dr. Garima

Dr. Bhavnesh Jaint

Department of Electrical Engineering

Delhi Technological University

Department of Electrical Engineering

Delhi Technological University

Date: May 2025

ABSTARCT

Analog signal processing involves manipulating continuous signals that vary in amplitude over time. This contrasts with digital signal processing, which deals with discrete signals shown by binary numbers. In analog processing, signals are manipulated using analog circuits, which can include components like resistors, capacitors, and operational amplifiers. Applications of analog signal processing can be found in various fields like audio processing, telecommunications, control systems, and instrumentation. This study investigates the design and implementation of analog circuits employing the second generation voltage conveyor (VCII+) as the core active building block. VCII+'s signify a notable advancement in analog signal processing, offering superior linearity, enhanced bandwidth, and improved power efficiency compared to conventional active elements such as operational amplifiers (op-amps). The study encompasses a comprehensive study of various analog circuit topologies, emphasizing the VCII+'s versatility and superior performance metrics. The proposed analog circuits are meticulously validated through simulations in Cadence PSpice, implemented in CMOS TSMC 0.18µm technology. The simulated characteristics closely align with the theoretical results. Furthermore, transient analysis, Monte Carlo simulations, temperature variation tests, and output noise evaluations are conducted to demonstrate the robustness and practical viability of the proposed analog circuit design.

CONTENTS

ACKNOWLEDGEMENT	ii
CANDIDATE'S DECLARATION	iii
CERTIFICATE BY THE SUPERVISOR	iv
ABSTRACT	v
LIST OF FIGURES	viii
LIST OF TABLES	X
1. INTRODUCTION	1-23
1.1 FUNDAMENTALS OF FILTER DESIGN:	1
1.2 TYPES OF FILTERS:	2
1.2.1 LOW-PASS FILTER (LPF):	3
1.2.2 HIGH-PASS FILTER (HPF):	5
1.2.3 BAND-PASS FILTER (BPF):	6
1.2.4 BAND-STOP FILTER (BSF) OR NOTCH FILTER: .	
1.2.5 ALL-PASS FILTER:	9
1.3 IMPORTANCE OF FILTERS:	10
1.4 ANALOG SIGNAL PROCESSING:	12
1.5 ACTIVE BUILDING BLOCKS:	14
1.6 FUNDAMENTALS OF VCII	16
1.7 VCII HISTORY:	19
1.8 VCII+ SYMBOLIC REPRESENTATION:	20
1.9 CMOS IMPLIMENTATIONS OF VCII+:	22
2. LITERATURE REVIEW	24-34
2.1 CIRCUIT TOPOLOGIES USING VCII AS ABB	24
2.2 COMPARATIVE STUDY	35
3. PROPOSED WORK AND SIMULATION RESULTS	36-42
3.1 REALIZATION OF ANALOG FILTERS USING PROPOSD C	IRCUIT:36

3.2 SIMULATION	AND EXPERIMENTAL RESULTS	OF PROPOSED
CIRCUIT:		38
4. FUTURE SCOPE	E AND CONCLUSION	43-44
4.1 FUTURE SCO	PE OF THE PRESENTED WORK:	43
4.2 CONCLUSION	NS:	44
5. REFERENCES		45-47

LIST OF FIGURES

FIGURE NO.	FIGURE TITLE	PAGE NO.
Figure 1.1	Characteristic curves of ideal filters	2
Figure 1.2	RC LPF Circuit	3
Figure 1.3	Frequency Response of LPF	3
Figure 1.4	RC HPF Circuit	5
Figure 1.5	Frequency Response of a HPF	5
Figure 1.6	RC BPF Circuit	7
Figure 1.7	Frequency Response of a BPF	7
Figure 1.8	Typical BSF Configuration	8
Figure 1.9	Frequency Response of a BSF	9
Figure 1.10	Frequency Response of APF	10
Figure 1.11	Phase Response of APF	10
Figure 1.12	VCII+ Symbolic Representation	21
Figure 1.13	Current Characteristics of VCII+	21
Figure 1.14	Voltage Characteristics of VCII+	22
Figure 1.15	CMOS Implementation of VCII+.	22
Figure 2.1	CMOS Implementation of VCII+ [1]	24
Figure 2.2	VCII as Voltage Amplifier [1]	25
Figure 2.3	VCII based Voltage Differentiator [1]	26
Figure 2.4	VCII based Voltage Integrator [1]	26
Figure 2.5	VCII as I to V Converter	27
Figure 2.6	VCII as V to I Converter	27
Figure 2.7	VCII as Current buffer	27
Figure 2.8	VCII as Voltage buffer	28
Figure 2.9	CMOS Implementation of VCII	28
Figure 2.10	VCII based first-order multifunction filter [2]	29
Figure 2.11	VCII based MISO voltage-mode filter circuit type 1 [2]	29
Figure 2.12	VCII based MISO voltage-mode filter circuit type 2 [2]	30
Figure 2.13	CMOS realization of VCII+ [3]	30
Figure 2.14	CMOS realization of VCII[3]	31
Figure 2.15	Equivalent Circuit [3]	31
Figure 2.16	VCII based APF [3]	31
Figure 2.17	CMOS implementation of VCII [4]	32
Figure 2.18	Third-order filter based on VCII [4]	33
Figure 2.19	Signal flow graph of RLC prototype low-pass filter [4]	33
Figure 2.20	Third-order RLC prototype [4]	34
Figure 3.1	Circuit of Realised Filter using VCII+	36

Figure 3.2	Theoretical and Simulated response of BPF	39
Figure 3.3	Theoretical and Simulated response of LPF	39
Figure 3.4	Theoretical and Simulated response of LPF	39
Figure 3.5	Theoretical and Simulated response of APF	40
Figure 3.6	Theoretical and Simulated response of BPF	40
Figure 3.7	Monte Carlo analysis of HPF	41
Figure 3.8	Monte Carlo analysis of LPF	41
Figure 3.9	Monte Carlo analysis of BPF	42

LIST OF TABLES

TABLE NO.	TITLE OF TABLE	PAGE No.
Table 1.1	Aspect Ratios of CMOS devices	23
Table 2.1	Comparison between various multifunction filers using VCII as an active element.	35

CHAPTER 1

1. INTRODUCTION

1.1 FUNDAMENTALS OF FILTER DESIGN:

Filter design is a fundamental aspect of electronic circuitry, playing a vital part in a multitude of applications like communication systems, signal processing, and instrumentation. The ability to manipulate signals through filtering operations is essential for achieving desired performance and functionality in electronic devices. Traditional filter design techniques often rely on conventional active and passive components, which may pose limitations in terms of complexity, power consumption, and performance. A filter is a circuit that processes signals on a frequency-dependent basis. Filters are categorized based on their magnitude response as low pass (LP), high pass (HP), band pass (BP), and band reject or notch (BR) filters. Another category includes AP filters, which manipulate phase while keeping the magnitude constant.

Filter design stands as a cornerstone in electronic circuitry, wielding significant influence across a spectrum of applications like communication systems, signal processing, and instrumentation. The ability to manage signals via filtering maneuvers is indispensable for achieving the desired efficacy and utility in electronic devices. Traditional methods of filter design often hinge on conventional utilization of both active and passive components. However, these approaches may encounter constraints concerning intricacy, power usage, and overall performance. In essence, a filter serves as a circuitry mechanism geared towards processing signals in accordance with their frequency attributes. Based on their response to magnitude, filters are categorized into distinct types, namely LP, HP, BP, AP and BR filters.

In the realm of electronic circuit design, filters serve as indispensable tools for signal processing, allowing engineers to manipulate signals according to specific frequency characteristics. While passive components like resistors, capacitors, and inductors traditionally have been used to construct filters, the emergence of active building blocks has revolutionized filter design by offering enhanced performance, flexibility, and efficiency. This chapter explores the principles, advantages, and methodologies involved in designing filters using active building blocks, shedding light on their significance in modern electronic systems. Filters constitute indispensable elements within electronic circuits used to selectively pass or attenuate signals based on their frequency content. They are utilized across a range of fields, spanning communication systems, audio manipulation, instrumentation, and control systems

1.2 TYPES OF FILTERS:

Filters are grouped according to their frequency response characteristics into various types:

- Low-pass filters (LPF): These filters enable signals below a given cutoff frequency to pass, and attenuating higher frequencies.
- High-pass filters (HPF): They permit signals above a certain cutoff frequency to pass through, and suppressing lower frequencies.
- Band-pass filters (BPF): These filters pass signals within a certain frequency range, while suppressing frequencies outside this range.
- Band-stop filters or notch filters (BSF): They obstruct signals within a certain frequency band, and permitting frequencies outside this band to pass.
- All-pass filters (APF): An APF is a signal processing tool that permits all
 frequencies to travel through it. However, its distinctive feature lies in its
 ability to modify the phase relationship between the input and output
 signals, while leaving the amplitude (or magnitude) of those frequencies
 unchanged.

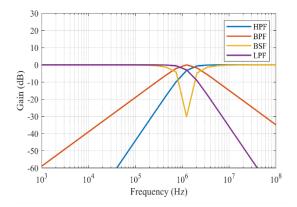


Figure 1.1: Characteristic curves of ideal filters

1.2.1 LOW-PASS FILTER (LPF):

Definition: A LPF is engineered to allow signals having frequencies lower than a designated cutoff frequency to pass through, while diminishing frequencies beyond this threshold. A LP filter can be constructed using a mix of capacitance, inductance, or resistance with the aim to generate substantial reduction above a designated frequency and minimal to no reduction below it. The frequency at which this transition happens is termed as "cut-off".

Application: LPFs are generally used in audio systems to remove high-frequency noise, in anti-aliasing filters for analog-to-digital converters, and in power supply circuits to eliminate high-frequency switching noise.

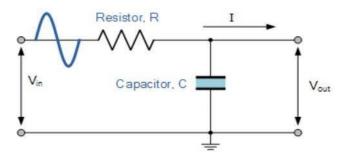


Figure 1.2: RC LPF Circuit

Implementation: LPFs can be constructed using passive components like resistors and capacitors in simple RC or RL configurations.

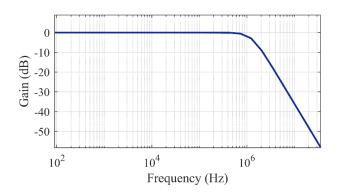


Figure 1.3: Frequency Response of LPF

By graphing the output voltage of the network versus various input frequencies, one can determine the Frequency Response or magnitude Bode Plot function of the LPF circuit, depicted below. A Bode plot for a LPF provides valuable insights into how the filter behaves across different frequencies. It visually represents the filter's frequency response, including its attenuation characteristics and phase shift behavior, making it a useful tool for analyzing and designing filter systems.

In a magnitude Bode plot, the vertical axis represents the magnitude of the system's response, typically measured in decibels (dB), and the horizontal axis represents frequency, usually on a logarithmic scale. For a LPF, the magnitude plot starts at 0 dB at low frequencies (close to DC) and decreases with increasing frequency. This represents the attenuation or suppression of higher frequencies by the filter. The slope of the magnitude plot in the low-frequency region is typically -20 dB/decade for a first order LPF. This means that for every tenfold increase in frequency, the magnitude decreases by 20 dB. For higher-order filters, the slope becomes steeper with each additional pole.

A Bode plot is a graphical representation of the frequency response of a system, which includes magnitude (in dB) and phase (in degrees) information as a function of frequency. Passive Low Pass Filters are commonly employed in various applications, particularly in audio amplifiers and speaker systems. Their primary function is to selectively allow lower frequency signals, such as bass, to pass through while attenuating higher frequency noise or distortions.

In audio applications, these filters are often referred to as "high-cut" filters. They work by employing a simple circuit configuration, typically consisting of a resistor and a capacitor arranged in series. The input signal is applied across this arrangement, with the output signal taken from across the capacitor.

The cut-off frequency, denoted as f_C , marks the point at which the filter begins to diminish the input signal. This frequency is expressed by the values of the resistor and capacitor based on the equation $f_C = 1/(2\pi RC)$. At the cut-off frequency, the output signal experiences a phase shift of -45 degrees, characteristic of a LPF.

1.2.2 HIGH-PASS FILTER (HPF):

Definition: A HPF allows signals with frequencies above a specified frequency to pass through while diminishing lower frequencies.

Application: HPFs are utilized in audio equalizers to boost high-frequency signals, in sensor applications to remove DC offset, and in crossover networks for speakers to separate low-frequency and high-frequency signals.

Implementation: HPFs can be made using passive components like capacitors and resistors in RC or RL configurations, or using active components such as op-amps in active filter designs.

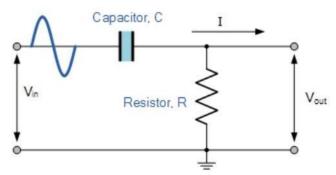


Figure 1.4: RC HPF Circuit

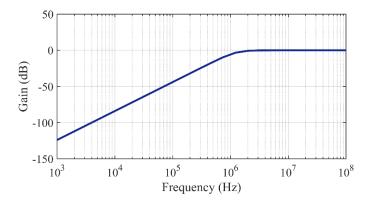


Figure 1.5: Frequency Response of a HPF

A Bode magnitude plot for a HPF provides a visual representation of its frequency response characteristics, including its attenuation behavior for low frequencies and phase shift properties. It's a valuable tool for analyzing and designing high-pass filters in various applications, such as signal processing, communications, and audio engineering. This plot provides a graphical representation of its frequency response in terms of magnitude (in dB) and phase (in degrees) as a function of frequency. For a HPF, the magnitude plot starts at 0 dB at low frequencies (close to DC) and increases with increasing frequency. This indicates that low-frequency components of the input signal are attenuated or suppressed by the filter. The slope of the magnitude plot in the low-frequency region is typically 0 dB/decade for a first- order HPF. This means that at low frequencies, the magnitude remains relatively constant.

The Bode magnitude Plot or Frequency Response Curve illustrated above for a passive HPF exhibits an inverse relationship compared to that of a LPF. Passive HPF find frequent application in audio amplifiers, often serving as coupling capacitors between amplifier stages. Within speaker systems, they redirect higher frequency signals to smaller "tweeter" speakers while suppressing lower bass signals. Additionally, they are utilized to reduce low-frequency noise or "rumble" distortion.

In the realm of audio, the HPF is occasionally referred to as a "low-cut" filter. When an AC sine wave is applied, it will act like a simple first-order high pass filter. However, if the input signal is changed to a pulse shape with an almost vertical step input, the circuit's response undergoes a significant change, resulting in a configuration commonly referred to as a differentiator.

1.2.3 BAND-PASS FILTER (BPF):

Definition: A BPF permits signals within a certain frequency range (bandwidth) to pass through while attenuating frequencies outside this range. Application: BPFs are essential in radio receivers for tuning to specific frequency bands, in medical devices for extracting physiological signals within a certain range, and in audio processing for isolating specific frequency components.

Implementation: BPFs can be realized using passive components like capacitors, inductors, and resistors in configurations such as LC filters, or using active components like op-amps in active filter designs.

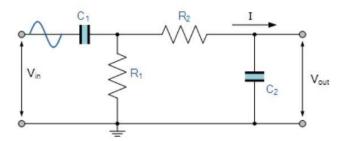


Figure 1.6: RC BPF Circuit

The Bode magnitude Plot, which illustrates the frequency response curve, reveals the behavior of the BPF. At lower frequencies, there's signal attenuation, while the output increases with a slope of +6dB/Octave until it reaches the "lower cut-off" frequency, marked as f_L . At this frequency, the output voltage reduces to 70.7% of the input signal value, corresponding to a - 3dB reduction. Subsequently, the output maintains maximum gain until it reaches the "upper cut-off" frequency, f_L beyond which it begins to decrease at a rate of -6dB/Octave, attenuating high-frequency signals.

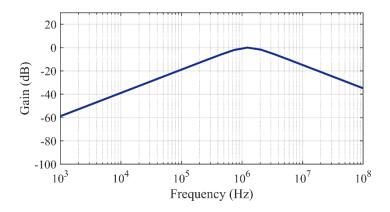


Figure 1.7: Frequency Response of a BPF

1.2.4 BAND-STOP FILTER (BSF) OR NOTCH FILTER:

Definition: A BSF, also known as a notch filter, diminishes signals within a certain frequency band while allowing frequencies outside this band to pass through.

Application: BSFs are used in audio systems to eliminate specific unwanted frequencies, in instrumentation to remove interference from power lines, and in biomedical devices for filtering out noise from physiological signals.

Implementation: BSFs can be implemented using passive components like capacitors, inductors, and resistors in configurations such as LC filters, or using active components like op-amps in active filter designs.

When combining a HP and LP filter, their frequency responses don't overlap as in a BP filter. This disparity arises because their starting and ending frequencies occur at different points. For instance, imagine a scenario where a first order LPF with a frequency f_L , of 200Hz, is linked in parallel with a 1st-order HPF having a frequency, f_L , of 800Hz. As these filters are effectively connected in parallel, the input signal undergoes, simultaneous processing by both filters.

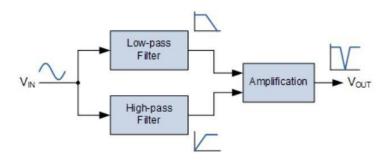


Figure 1.8: Typical BSF Configuration

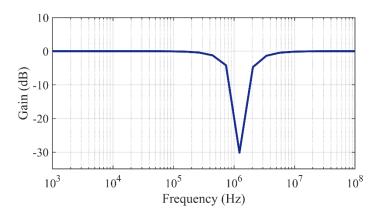


Figure 1.9: Frequency Response of a BSF

Frequencies below 200Hz traverse the LPF without attenuation, while frequencies above 800Hz pass through the HPF without alteration. However, frequencies within the range of 200Hz to 800Hz are rejected by either filter, resulting in a notch in the filter's output response. Essentially, signals with frequencies of 200Hz or lower and 800Hz or higher pass through unaffected, whereas signals with frequencies between these ranges, such as 500Hz, are rejected as they are too high for the LPF and too low for the HPF to transmit. Upon examining the amplitude and phase curves for the BS circuit, it becomes evident that the parameters fL, fH, and fC closely resemble those utilized in describing the behavior of the BSF.

1.2.5 ALL-PASS FILTER:

Definition: An APF is designed to pass all frequencies with equal gain but introduces a phase shift that varies with frequency.

Application: They are used in audio processing for phase correction, in equalization circuits for time delay correction, and in communication systems for signal synchronization.

Implementation: These filters can be constructed using passive components like capacitors, inductors, and resistors, or using active components such as op-amps. All pass filters are characterized by a consistent frequency response, meaning they neither accentuate nor diminish any specific part of the spectrum. Instead, they alter the timing of signals based on their frequency content. This temporal adjustment, dictated by the phase response, is the hallmark of APF.

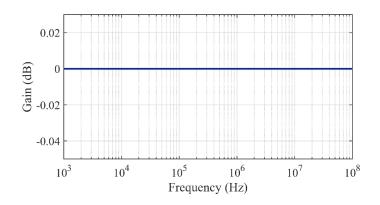


Figure 1.10: Frequency Response of APF

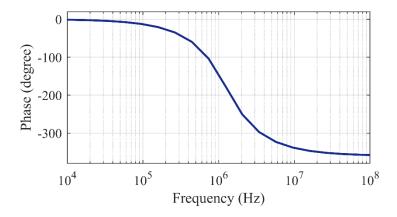


Figure 1.11: Phase Response of APF

In circuit design, all-pass filters serve diverse functions that rely on frequency-dependent time alignment. They find applications in various audio contexts such as filter banks, speaker crossovers, and reverberators. Additionally, these filters are utilized in both continuous-time and discrete-time scenarios.

1.3 IMPORTANCE OF FILTERS:

Filters are integral components in various electronic systems and find applications in:

 Communication Systems: Filtering out unwanted noise and interference, channel selection, and equalization in wireless, wired, and optical communication systems.

- Audio Systems: Equalization, tone control, crossover networks, and noise reduction in audio amplifiers, mixers, and speakers.
- Instrumentation and Measurement: Signal conditioning, anti-aliasing filtering, and noise rejection in data acquisition systems, sensors, and test instruments.
- Biomedical Devices: Signal filtering and processing in medical imaging, patient monitoring, and diagnostic equipment.
- Control Systems: Filtering noise and disturbances, signal shaping, and feedback control in industrial automation, robotics, and automotive systems.

Filter design involves selecting appropriate components and configurations to achieve the desired frequency response. The basic principles of filter design include:

- Frequency Response: Describes how a filter affects the amplitude and phase of signals at different frequencies.
- Cutoff Frequency: The frequency at which the filter's response starts to change significantly.
- Order: Indicates the complexity of the filter and determines its roll-off rate and selectivity.
- Transfer Function: Mathematical representation of the filter's input-output relationship.

Filters can be implemented using passive components (resistors, capacitors, inductors) or active components (op-amps, transistors, etc.). Common filter configurations include:

- Passive RC Filters: Simple filters using resistors and capacitors to achieve basic filtering functions.
- Active Filters: Employ active components like operational amplifiers to achieve more complex filtering operations with gain and bandwidth control.
- Digital Filters: Implemented using DSP techniques in software or hardware, offering precise control and flexibility.

Filters perform an important role in building the behavior of electronic circuits, enabling precise control over signal characteristics and facilitating a wide range of applications across different industries. As technology

advances, filters continue to evolve, offering improved performance, efficiency, and versatility in electronic systems. The history of filter designing reflects a continuous quest for improved performance, efficiency, and versatility in signal processing. From the early concepts of passive filtering to the sophisticated algorithms of modern digital filters, filter design has been instrumental in shaping the evolution of telecommunications, electronics, and information technology.

1.4 ANALOG SIGNAL PROCESSING:

Analog signal processing involves manipulating continuous signals that vary in amplitude over time. This contrasts with digital signal processing, which deals with discrete signals shown by binary numbers. In analog processing, signals are manipulated using analog circuits, which can include components like resistors, capacitors, and operational amplifiers. Applications of analog signal processing can be found in various fields like audio processing, telecommunications, control systems, and instrumentation. Common tasks include filtering, amplification, modulation, demodulation, and signal conditioning. Analog signal processing is often used when the input or output signals are naturally analog, or when the cost or complexity of implementing digital processing is prohibitive. However, digital signal processing has become increasingly dominant due to its flexibility, accuracy, and the advancement of digital technology.

Key operations in analog signal processing (ASP) include:

- 1. Filtering: Filtering involves modifying the frequency content of a signal. This can be achieved through passive components like resistors and capacitors in configurations such as RC filters, or through active components like operational amplifiers in active filter designs.
- 2. Amplification: Amplification increases the magnitude of a signal. Opamps are commonly used for this purpose, configured in various amplifier configurations such as voltage amplifiers, current amplifiers, and transconductance amplifiers.

- 3. Modulation and Demodulation: Modulation encompasses impressing information onto a carrier signal, and demodulation is the process of extracting this information from the modulated signal. Techniques like AM, FM, PM are employed for communication purposes.
- 4. Signal Conditioning: Signal conditioning involves preparing a signal for further processing or transmission. This may include tasks such as impedance matching, noise reduction, level shifting, and signal isolation.

Analog signal processing finds applications in numerous domains including audio processing, telecommunications, instrumentation, control systems, and sensor interfacing. While digital signal processing has gained prominence due to its versatility and computational efficiency, analog signal processing remains indispensable in scenarios where real-world signals are inherently analog or where strict requirements on signal fidelity, speed, or power consumption dictate the use of analog techniques.

Utilizing active building blocks for filter design presents a dynamic approach in electronic circuitry, providing a rich array of options for signal manipulation and frequency control. LPF, employed to diminish high-frequency signals while permitting low-frequency ones, can be effectively crafted using active components like operational amplifiers (op-amps) in tandem with resistors and capacitors. Conversely, HPF, functioning to pass high-frequency signals while suppressing lower frequencies, can capitalize on active building blocks to achieve precise cutoff frequencies and enhance roll-off characteristics. Bandpass filters, crucial for isolating signals within specific frequency bands, can take shape through active components in configurations such as multiple feedback (MFB) or state-variable filters, offering adjustable center frequencies and bandwidths. Similarly, band-stop filters, known for attenuating signals within designated frequency bands while transmitting frequencies beyond, can be efficiently realized with active components to improve selectivity and eliminate unwanted frequencies. Furthermore, active building blocks facilitate the design of intricate filter structures like multiple-order filters, elliptic filters, and switched capacitor filters, boasting advanced functionalities such as rapid roll-off, precise transition bands, and customizable cutoff frequencies. Ultimately, the adaptability and versatility of active building blocks empower engineers to tailor filter designs precisely to meet the unique requirements of diverse applications with efficiency and accuracy.

Active building blocks play a very important role in electronic circuit design, offering a versatile and efficient platform for signal processing, amplification, and control. With advancements in technology and design methodologies, active building blocks continue to drive innovation across various industries, enabling the development of advanced electronic systems with enhanced performance and functionality.

1.5 ACTIVE BUILDING BLOCKS:

Active building blocks refer to electronic components or modules that incorporate active devices such as op-amps, transistors, or voltage/current sources to provide amplification, signal conditioning, or processing capabilities. These building blocks offer distinct advantages over passive components, including:

- 1. Gain: Active components can provide signal amplification, enabling the design of filters with higher gain levels compared to passive filters.
- 2. Flexibility: Active building blocks allow for the implementation of complex filter configurations and frequency responses that may be challenging to achieve using passive components alone.
- 3. Tunability: Active filters provide the flexibility to modify parameters such as cutoff frequency, BW, and gain by adjusting component values or control inputs. This enables easy adaptation to different system demands.
- 4. Low Sensitivity: Active filters are less sensitive to component variations and parasitic effects, resulting in improved stability and robustness.
- 5. Integration: Active building blocks can be readily integrated into integrated circuit (IC) technologies, enabling compact and cost-effective solutions for filter design.

The design of various filters using active building blocks typically involves the following steps:

- 1. Specification: Define the desired filter specifications including frequency response, cutoff frequency, bandwidth, and gain requirements.
- 2. Topology Selection: Choose a suitable filter topology based on the specified requirements and application constraints. Common filter types include LP, HP, BP and BS.
- 3. Component Sizing: Determine the component values (resistors, capacitors, etc.) and operating parameters of active devices (op-amps, transistors)

- based on the selected topology and specifications.
- 4. Simulation and Optimization: Utilize circuit simulation tools to verify the performance of the designed filter and optimize component values for desired characteristics such as frequency response, passband ripple, and stopband attenuation.
- 5. Sensitivity Analysis: Conduct sensitivity analysis to examine the impact of component tolerances, temperature variations, and manufacturing variations on filter performance and stability.
- 6. Prototype Implementation: Build a physical prototype of the designed filter circuit using discrete components or integrated circuits, following best practices for layout and grounding to minimize noise and interference.
- 7. Performance Evaluation: Test the prototype filter under various operating conditions to evaluate its performance in terms of frequency response, gain, noise, distortion, and stability.
- 8. Fine-Tuning and Adjustment: Fine-tune the filter circuit parameters if necessary to meet any deviations from the desired specifications, taking into account practical constraints and limitations.

Filters designed using active building blocks find wide-ranging applications across different domains, including:

- 1. Communication Systems: Active filters are used for frequency shaping, channel selection, and signal conditioning in communication systems such as wireless transceivers, satellite receivers, and base stations.
- 2. Audio Processing: Active filters are employed in audio equalizers, crossover networks, and tone control circuits for audio processing applications in amplifiers, mixing consoles, and audio recording equipment.
- 3. Biomedical Instruments: Active filters are utilized in biomedical instrumentation for signal filtering and noise rejection in electrocardiography (ECG), electroencephalography (EEG), and other medical diagnostic systems.
- 4. Instrumentation and Control: Active filters play a very important role in instrumentation and control systems for signal conditioning, anti-aliasing filtering, and noise reduction in sensors, data acquisition systems, and industrial automation equipment.
- 5. Automotive Electronics: Active filters are integrated into automotive electronic systems for audio entertainment, engine control, and safety features such as active noise cancellation and adaptive cruise control.
- 6. Consumer Electronics: Active filters are incorporated into consumer electronic devices such as televisions, smartphones, and portable audio

players for audio processing, speaker protection, and noise filtering.

Recent advancements in active building blocks have focused on enhancing performance, efficiency, and integration. These advancements include:

- 1. High-Speed Op-amps: Op-amps with increased bandwidth and slew rate enable high-speed signal processing in communication and data acquisition systems.
- 2. Low-Power ICs: Low-power integrated circuits conserve energy and extend battery life in portable electronic devices and IoT applications.
- 3. Integrated Sensor Interfaces: ICs incorporating sensor interfaces simplify the integration of sensors into electronic systems, enabling IoT, healthcare, and environmental monitoring applications.
- 4. Programmable Filters: Programmable filter ICs offer configurable filter characteristics, allowing for flexible signal processing in audio, communication, and instrumentation applications.

1.6 FUNDAMENTALS OF VCII

The Second Generation Voltage Conveyor (VCII) is a versatile and essential analog building block in modern circuit design. It bridges the gap between voltage and current domains, facilitating efficient signal processing in analog systems. Unlike operational amplifiers, which dominate classical analog design, VCII is specifically engineered to overcome limitations related to bandwidth, power consumption, and linearity. The growing complexity and demands of analog circuits in applications like telecommunications, instrumentation, signal filtering, and biomedical devices have made VCII an indispensable component in the toolkit of circuit designers.

VCII is a three-port active circuit element that operates based on the principle of voltage- controlled current transfer. Its design ensures a unique relationship among its three terminals: Y, X, and Z. The Y port serves as the input voltage port, the X port is a high-impedance node controlled by the voltage applied to Y, and the Z port is the output port providing current proportional to the current at X. This structure facilitates the direct processing and transformation of signals between voltage and current forms, which is vital in numerous analog signal-processing tasks.

The importance of VCII stems from its ability to maintain high linearity and wide bandwidth, even under challenging operating conditions. Traditional analog devices like operational amplifiers struggle with limitations such as reduced performance at high frequencies and higher power consumption in certain applications. VCII addresses these limitations by providing superior high-frequency characteristics and

low power dissipation. These advantages make it an ideal choice for designing modern analog circuits that require efficiency and reliability.

One of the primary reasons VCII gained popularity is its compatibility with complementary metal-oxide-semiconductor (CMOS) technology. CMOS technology, known for its low power consumption and scalability, provides an effective platform for implementing VCII in integrated circuits (ICs). This compatibility has allowed VCII to be utilized in compact and power-efficient designs, further expanding its application range. As a result, VCII has found extensive use in signal filtering, impedance matching, oscillator circuits, and other critical analog circuit functions.

Historically, the development of VCII can be traced back to the limitations encountered in first-generation voltage conveyors (VCI). While VCI served as a precursor to VCII, it lacked the precision and frequency response necessary for advanced applications. The introduction of VCII addressed these issues by incorporating improved design features such as higher input impedance, better current transfer accuracy, and enhanced high-frequency performance. These improvements marked a significant milestone in the evolution of voltage conveyor technology, allowing VCII to become a cornerstone of modern analog design.

The operation of VCII is governed by straightforward behavioral equations. The voltage at the intermediate X port is equal to the voltage applied at the Y input port, while the current at the Z output port mirrors the current at the X port. These fundamental relationships form the basis for its versatility and ease of integration into analog systems. Moreover, the simplicity of its operation equations ensures that VCII can be easily modeled and simulated using modern circuit simulation tools such as PSPICE, LTSPICE, and MATLAB.

In practical applications, VCII's unique characteristics enable it to perform functions such as signal amplification, voltage-to-current conversion, and impedance scaling. These functions are crucial in designing active filters, oscillators, and other analog circuits where precise signal control is required. For instance, VCII can be used to create high-performance filters that offer improved selectivity and reduced component count compared to traditional designs. Similarly, in oscillator circuits, VCII contributes to stable and efficient frequency generation, a critical requirement in communication systems and instrumentation.

VCII's role is not limited to theoretical designs; it has demonstrated its effectiveness in real- world applications. Industries such as telecommunications rely on VCII for the development of robust and efficient circuits for signal transmission and processing. Similarly, in biomedical devices, VCII is employed in signal conditioning circuits that handle bioelectrical signals with high precision. Its versatility also extends to control systems, where it plays a role in designing feedback loops and compensators.

The continuous advancement in CMOS technology has further enhanced the capabilities of VCII, making it suitable for next-generation applications. With the increasing demand for high-speed and low-power analog systems, VCII remains a preferred choice for designers aiming to balance performance and efficiency. Its ability to operate in a wide range of voltage and frequency conditions ensures that it can meet the diverse requirements of modern analog circuits.

In conclusion, VCII represents a significant advancement in the field of analog circuit design. Its superior performance characteristics, coupled with its compatibility with CMOS technology, have established it as a foundational component for a wide array of applications. From signal processing and amplification to oscillation and filtering, VCII continues to drive innovation in analog electronics, making it an indispensable tool for researchers and designers. Its historical evolution, operational principles, and practical significance underscore its pivotal role in shaping the future of analog circuit design.

1.7 VCII HISTORY:

The history of the Second Generation Voltage Conveyor (VCII) is closely tied to the evolution of analog circuit design and the need for efficient signal processing components. As electronic systems advanced during the latter half of the 20th century, designers faced challenges in creating high-performance circuits capable of handling increasingly complex requirements. These challenges led to the development of innovative active elements, including the concept of voltage conveyors. The introduction of VCII was a direct response to the limitations observed in earlier designs, marking a significant milestone in the progress of analog electronics.

The concept of voltage conveyors was first introduced in the 1960s and 1970s. Sedra and Smith formalized the idea of current conveyors, which were the precursors to voltage conveyors, in their seminal work. These devices offered a new approach to analog circuit design by providing efficient current-mode operation, a contrast to the voltage-mode operation of traditional operational amplifiers. While current conveyors gained traction for their ability to handle high-frequency signals with reduced parasitic effects, they were not without limitations. The first-generation voltage conveyor (VCI), developed as an extension of the current conveyor, struggled with performance issues, particularly in terms of frequency response and linearity. These challenges created the demand for an improved version, leading to the development of VCII.

The design of VCII emerged during a time when advancements in semiconductor technology were enabling the miniaturization and integration of circuit components. Researchers and engineers recognized the need for a more versatile and efficient voltage conveyor that could address the shortcomings of VCI. VCII was conceived to provide enhanced performance characteristics, including higher accuracy, better bandwidth, and improved compatibility with complementary metal-oxide-semiconductor (CMOS) technology. Its development was driven by the need to create active elements that could meet the demands of emerging applications in telecommunications, instrumentation, and signal processing.

The evolution of VCII was heavily influenced by the progress in CMOS technology, which played a pivotal role in its practical implementation. By the 1980s, CMOS technology had become the standard for integrated circuit design due to its low power consumption and scalability. This compatibility allowed VCII to be effectively realized in integrated circuits, making it accessible for a wide range of applications. The adoption of CMOS-based VCII designs enabled the development of compact and power-efficient analog circuits, further solidifying its

role in modern electronics.

As the second-generation voltage conveyor, VCII introduced several critical improvements over its predecessor. Its high input impedance and low output impedance ensured minimal signal loss and distortion, which were significant limitations of VCI. Additionally, VCII's ability to transfer signals accurately across a wide frequency range made it particularly valuable in high-speed applications. These enhancements addressed the needs of designers who required reliable and precise components for tasks such as signal filtering, impedance matching, and oscillation.

The historical significance of VCII extends beyond its technical advancements. It represents a shift in analog circuit design philosophy, moving away from the dominance of operational amplifiers and towards specialized active elements tailored for specific functions. This shift was instrumental in enabling the design of more efficient and versatile analog systems, laying the foundation for many of the high-performance circuits in use today. The development of VCII also exemplifies the iterative nature of technological progress, where the limitations of one generation of devices drive the innovation and refinement of the next.

In recent decades, VCII has continued to evolve alongside advancements in semiconductor technology. Modern VCII implementations leverage cutting-edge processes to achieve even greater performance metrics, such as higher linearity, wider bandwidth, and lower power consumption. These improvements have expanded the range of applications for VCII, from telecommunications and industrial automation to biomedical instrumentation and consumer electronics. The enduring relevance of VCII underscores its importance as a foundational component in the field of analog design.

In summary, the history of VCII is a testament to the ongoing quest for innovation in electronic systems. Its development was driven by the need to overcome the limitations of earlier designs, and its success has been fueled by continuous advancements in technology. As an active element that combines efficiency, versatility, and precision, VCII remains a cornerstone of analog circuit design, reflecting both the progress of the past and the potential for future breakthroughs.

1.8 VCII+ SYMBOLIC REPRESENTATION:

The symbolic representation of the Second Generation Voltage Conveyor (VCII+) provides an intuitive way to understand its functionality and integration into analog circuits. VCII+ is

a three-port device with terminals Y, X, and Z, each serving specific roles. The X port replicates the voltage from Y, while the Z port provides an output current equal to the current at X. These behaviors are captured by the relationships $V_Z=V_X$ and $I_X=I_Y$. Represented as a simple block diagram, this symbolic depiction abstracts the internal design, focusing on its operational relationships, making it easier for designers to visualize and incorporate VCII into various circuits.

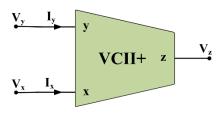


Figure 1.12: VCII+ Symbolic Representation

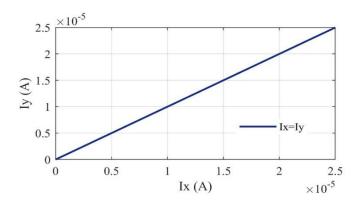


Figure 1.13: Current Characteristics of VCII+

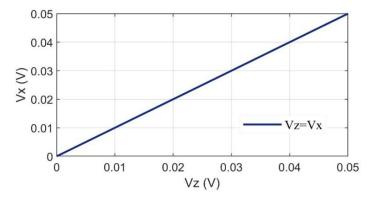


Figure 1.14: Voltage Characteristics of VCII+

1.9 CMOS IMPLIMENTATIONS OF VCII+:

The CMOS implementation of the Second Generation Voltage Conveyor (VCII+) leverages the advantages of CMOS technology, such as low power consumption and scalability, to achieve its functional characteristics of accurate voltage and current transfer

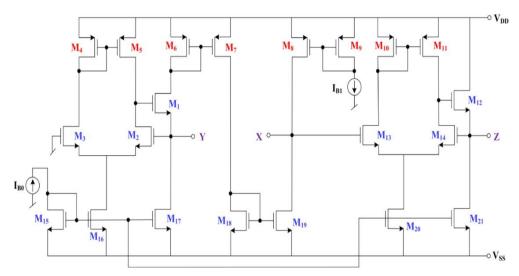


Figure 1.15: CMOS Implementation of VCII+.

Challenges such as power-speed trade-offs and scaling-induced variability are addressed through optimized transistor sizing, biasing, and layout strategies. CMOS implementation allows VCII+ to integrate seamlessly into mixed-signal systems and SoCs, enabling compact and efficient designs for applications like signal processing and telecommunications. This adaptability ensures that VCII+ remains relevant in modern analog circuit design, leveraging CMOS advancements for improved performance and reliability.

To check how well the proposed grounded parallel immittance simulator circuits work, we used CMOS VCII+ designs. These were built using 0.18- μ m TSMC technology, as shown in Figure 1.15. The sizes of the PMOS and NMOS transistors used in Figure 1.15 are listed in Table 1.1. VCII+ was powered by ± 0.9 V supply voltages, and the bias currents IB₀ and IB₁ were set to 25 μ A.

Table 1.1: Aspect Ratios of CMOS devices

CMOS Transistor	W(µm)	L(µm)
M1 - M3 N	13.5	0.54
M4 - M11 P	40.5	0.54
M12 - M21 N	13.5	0.54

CHAPTER 2

2. LITERATURE REVIEW

The **Second-Generation Voltage Conveyor (VCII)** has emerged as a key component in analog signal processing due to its superior capabilities over earlier active building blocks (ABBs) like the Second-Generation Current Conveyor (CCII). Its ability to handle voltage- mode and current-mode signals with high efficiency has made it indispensable in the design of advanced filters, oscillators, and other signal processing circuits.

2.1 CIRCUIT TOPOLOGIES USING VCII AS ABB

The CMOS implementation of a VCII+ as proposed by [1] is shown in Figure 2.1. This circuit consists of two main blocks: a current buffer between the Y and X terminals and a voltage buffer between the X and Z terminals. • The current buffer is formed using transistors M1–M7 and current sources IB1–IB4. • The voltage buffer includes M8, MA1–MA3 and current sources IB5–IB7.

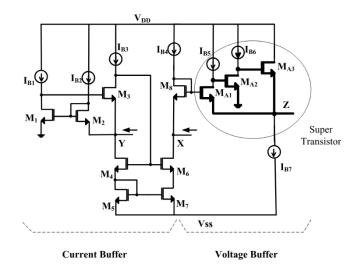


Figure 2.1: CMOS Implementation of VCII+ [1]

Negative feedback loops are used to control the impedance and improve accuracy. For instance, the loop formed by M1–M3 ensures the Y terminal is grounded and has low impedance. Another loop using M4–M7 transfers the input current to terminal X. The outputs of its three ports in terms of their corresponding inputs at relatively low frequencies are represented by the following matrix equation:

$$\begin{bmatrix} I_x \\ V_z \\ V_Y \end{bmatrix} = \begin{bmatrix} \pm \beta & 0 & 0 \\ 0 & \alpha & 0 \\ 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} I_y \\ V_x \\ I_z \end{bmatrix}$$
 (2)

 β is the current gain between Y and X terminals

If $\beta = +1$, we are considering VCII+

If $\beta = -1$, we are considering VCII-

α is the voltage gain between X and Z terminals

VCII-Based Applications

The paper demonstrates VCII's use in various analog functions like voltage amplification, differentiation, integration, I-to-V and V-to-I conversion.

Voltage Amplifier

The gain is given by:

$$A_V = \mp \beta \alpha \cdot \frac{R_2}{R_1} \tag{3}$$

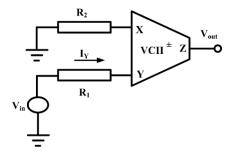


Figure 2.2: VCII as Voltage Amplifier [1]

Voltage Differentiator & Integrator

Differentiator TF is given by:

$$\frac{V_{out}}{V_{in}} = \mp \beta \alpha \cdot \text{sCR}$$
 (4)

Integrator TF is given by:

$$\frac{V_{out}}{V_{in}} = \mp \frac{\beta \alpha}{\text{sCR}}$$
 (5)

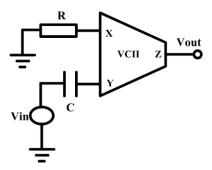


Figure 2.3: VCII based Voltage Differentiator [1]

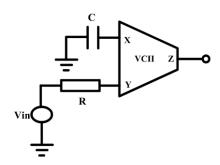


Figure 2.4: VCII based Voltage Integrator [1]

I-to-V and V-to-I Converters

I-to-V TF is given by:

$$\frac{V_{out}}{I_{in}} = \mp \beta \alpha \cdot R \tag{6}$$

.

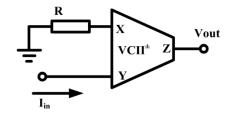


Figure 2.5: VCII as I to V Converter

V-to-I TF is given by:

$$\frac{I_{out}}{V_{in}} = \mp \frac{\beta}{R} \tag{7}$$

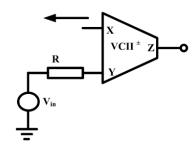


Figure 2.6: VCII as V to I Converter

Voltage and Current Buffers

Current buffer ($\beta = 1$):

$$\frac{I_{out}}{I_{in}} = 1 \tag{8}$$

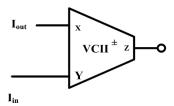


Figure 2.7: VCII as Current buffer

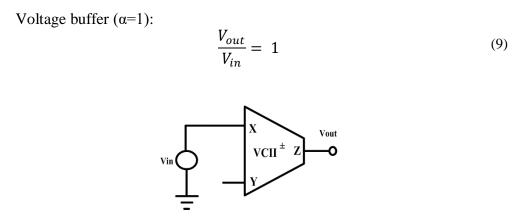


Figure 2.8: VCII as Voltage buffer

The CMOS implementation of the second-generation voltage conveyor (VCII) as proposed by [2] is shown in Figure 2.9. This structure uses a low-voltage, low-power design that combines two unity-gain buffers: a current buffer (Y to X) and a voltage buffer (X to Z). The current buffer, composed of transistors M1–M4 and M9–M12, provides the unity gain current transfer ($I_X = I_Y$). The voltage buffer, made up of M5–M7 and M13–M15, ensures unity gain voltage transfer ($V_Z = V_X$). Both buffers utilize two-stage op-amp configurations operating in closed-loop mode.

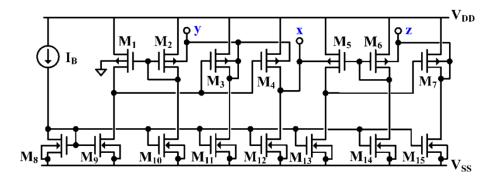


Figure 2.9: CMOS Implementation of VCII

The multifunction filter is constructed using one VCII and three passive elements: Z1, Z2, and Z3, as shown in Figure 2.10. Vin1 and Vin2 are input voltage nodes; Vo is the output node at terminal Z. The configuration allows different filter types by assigning values to Z1–Z3 and applying input signals accordingly.

The general output voltage equation is given as:

$$V_o = \frac{Z_1 Z_3 V_{in2} - Z_2 Z_3 V_{in1}}{Z_1 (Z_2 + Z_3)} \tag{10}$$

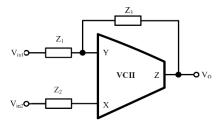


Figure 2.10: VCII based first-order multifunction filter [2]

Filter Type 1

For Type 1 configuration (shown in Figure 2.11), $Z_1 = R_1$, $Z_2 = C$, and $Z_3 = R_f$. The filter achieves three standard responses: low-pass, high-pass, and all-pass. The output voltage is given by:

$$V_o = \frac{\text{sC}R_f V_{in2} - (R_f/R_1)V_{in1}}{(\text{sC}R_f + 1)}$$
(11)

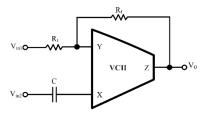


Figure 2.11: VCII based MISO voltage-mode filter circuit type 1 [2]

Filter Type 2

For Type 2 configuration (Figure 2.12), $Z_1 = C_1$, $Z_2 = R$, and $Z_3 = C_2$. This also allows LPF, HPF, and APF responses, with the output voltage expressed as:

$$V_o = \frac{V_{in2} - sC_1RV_{in1}}{(sCR_f + 1)}$$
(12)

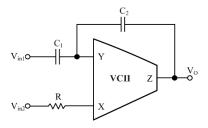


Figure 2.12: VCII based MISO voltage-mode filter circuit type 2 [2]

The CMOS implementation of the second-generation voltage conveyor (VCII) proposed by [3] is provided in Figure 2.13 and 2.14. It includes two variants: VCII+ and VCII-. Both designs are optimized for low-voltage operation and employ standard TSMC 0.18 μ m technology. PMOS transistors use a W/L ratio of 9 μ m/0.9 μ m, and NMOS transistors use 27 μ m/0.9 μ m. The circuit is biased with $\pm 1.2V$ and a bias current IB = 20 μ A.

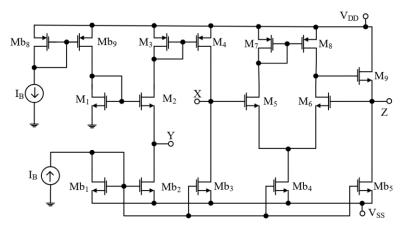


Figure 2.13: CMOS realization of VCII+.[3]

The equivalent circuit are shown in Figure 2.15. The fundamental relation among its terminals is given by the matrix equation:

$$\begin{bmatrix} i_x \\ V_z \end{bmatrix} = \begin{bmatrix} \pm \beta & 0 \\ 0 & \alpha \end{bmatrix} \begin{bmatrix} i_y \\ V_x \end{bmatrix} \tag{13}$$

Here, β is the current transfer ratio from terminal Y to X, and α is the voltage transfer ratio from terminal X to Z. Depending on the sign of β , either VCII+ or VCII- is realized.

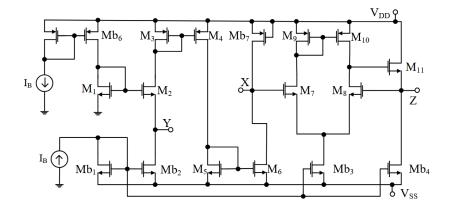


Figure 2.14: CMOS realization of VCII- [3]

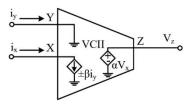


Figure 2.13: Equivalent Circuit [3]

The all-pass filter (APF) circuit is shown in Figure 2.16. It consists of three VCIIs, five resistors, and one grounded capacitor. The filter operates in transimpedance mode, receiving a current input and producing a phase-shifted voltage output. It is designed with low power, high linearity, and good IC compatibility.

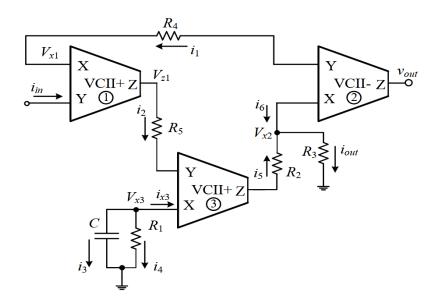


Figure 2.14: VCII based APF [3]

Filter Operation and Equations

Using the VCII properties and standard nodal analysis on the filter in Figure 2.16, the following transfer function is derived (assuming $R_4 = R_5$):

$$\frac{V_{out}}{I_{in}} = R_3 \frac{(-R_1/R_2 + \text{sC}R_1 + 1)}{1 + \text{sC}R_1}$$
(14)

Letting $R_1 = 2R_2$, the equation simplifies to:

$$\frac{V_{out}}{I_{in}} = \frac{R_3(sCR_1 - 1)}{1 + sCR_1}$$
 (15)

Passband Gain:

$$\frac{V_{out}}{I_{in}} = R_3 \tag{16}$$

Phase shift (ϕ) as a function of R1 and C is given by:

$$\varphi = \pi - 2\tan^{-1}(\omega R_1 C) \tag{17}$$

Pole frequency:

$$\omega p = \frac{1}{R_1 C} \tag{18}$$

The CMOS implementation of VCII proposed by [4] is shown in Figure 2.17. It is constructed using two unity-gain op-amps, one for current transfer (Y to X) and the other for voltage transfer (X to Z). The structure utilizes bulk-driven MOS transistors operating in the subthreshold region, enabling ultra-low power operation at a supply voltage of 0.5 V.

- Current follower (Ix = Iy): Implemented using M1–M4 and M9–M12.
- Voltage follower (Vz = Vx): Implemented using M5–M7 and M13–M15.
- Biasing is achieved using transistor M8 and a bias current IB.

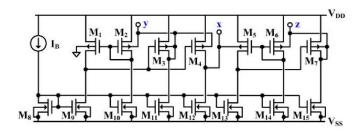


Figure 2.15: CMOS implementation of VCII [4]

The output resistances of the y, x and z terminals can be calculated as:

$$r_{y} \approx (g_{oM1} + g_{oM6})/(g_{mM3} \cdot g_{mbM1}) \tag{19}$$

$$r_{x} = 1/(g_{0M4} + g_{0M12}) \tag{20}$$

$$r_z \approx (g_{oM5} + g_{oM13})/(g_{mM7} \cdot g_{mbM6}) \tag{21}$$

where gm, gmb, go are the transconductance, bulk transconductance and output conductance of the MOS transistor, respectively

The filter is designed using three cascaded VCII-based integrator stages that replicate the RLC ladder prototype. The filter can operate in current-mode (low input and high output impedance) or transimpedance-mode (low input and output impedance).

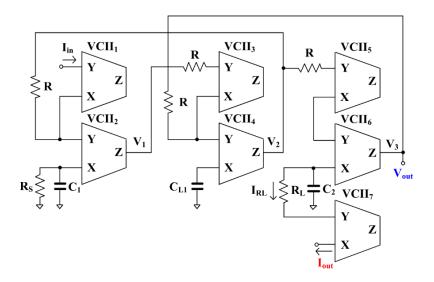


Figure 2.16: Third-order filter based on VCII [4]

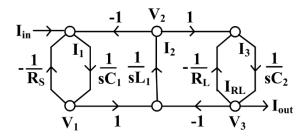


Figure 2.17: Signal flow graph of RLC prototype low-pass filter [4]

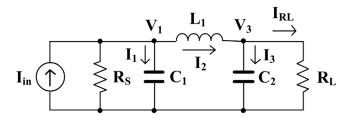


Figure 2.18: Third-order RLC prototype [4]

Filter Equations

Filter Equations
$$\begin{split} I_1 &= I_{in} - V_1/R_s - I_2 & (22) \\ V_1 &= I_1 \ / \ (sC_1) & (23) \\ I_2 &= \ (V_1 - V_2) \ / \ (sL_2) & (24) \\ V_3 &= \ (I_2 - I_{RL}) \ / \ (sC_2) & (25) \\ I_3 &= I_2 - I_{RL}, & (26) \\ \text{where } I_{RL} &= I_{out}, \ V_3 = V_{out} \end{split}$$

2.2 COMPARATIVE STUDY

Table 2.1: Comparison between various multifunction filers using VCII as an active element.

Ref	Mode	No. of VCII	Technology	No. of R+C	All grounded passive element only	Filtering functions
[2]	VM	1	0.18 μm CMOS	2+1	No	LP, AP, HP
	VM	1	0.18 μm CMOS	1+2	No	LP, AP, HP
[3]	TIM	3	0.18 μm CMOS	5+1	No	APF
[4]	CM	7	0.18 μm CMOS	6+3	No	LP
Proposed Circuit	VM	3	0.18 μm CMOS	6+2	NO	LP, HP, BP BR, AP

CHAPTER 3

3. PROPOSED WORK AND SIMULATION RESULTS

3.1 REALIZATION OF ANALOG FILTERS USING PROPOSD CIRCUIT:

The proposed filter used three VCII+ and eight passive components (two capacitor and six resistor) is demonstrated in Fig.3.1. The transfer function of the proposed filter is given by:

$$V_o = \frac{s^2 V_{in2} + s \frac{1}{R_3 C_2} V_{in3} + \frac{K}{C_1 C_2 R_1 R_2} V_{in1}}{s^2 + s \frac{1}{R_3 C_2} + \frac{K}{C_1 C_2 R_1 R_2}}$$
(27)

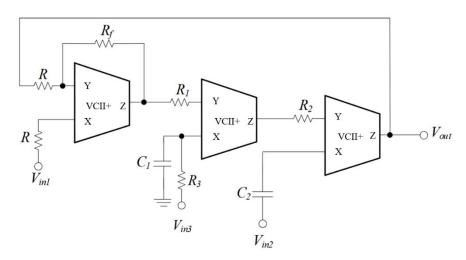


Figure 3.1: Circuit of Realized Filter using VCII+

Depending upon the voltage status of V_1 , V_2 and V_3 in the above equation one of the following five filter functions is realized.

1. For LPF, $V_3 = V_2 = 0$ and $V_1 = V_{in}$;

$$\frac{V_o}{V_{in}} = \frac{\frac{K}{C_1 C_2 R_1 R_2}}{s^2 + s \frac{1}{R_3 C_2} + \frac{K}{C_1 C_2 R_1 R_2}}$$
(28)

2. For HPF, $V_3 = V_1 = 0$ and $V_2 = V_{in}$;

$$\frac{V_o}{V_{in}} = \frac{s^2}{s^2 + s \frac{1}{R_3 C_2} + \frac{K}{C_1 C_2 R_1 R_2}}$$
(29)

3. For BPF, $V_2 = V_1 = 0$ and $V_3 = V_{in}$;

$$\frac{V_o}{V_{in}} = \frac{s \frac{1}{R_3 C_2}}{s^2 + s \frac{1}{R_3 C_2} + \frac{K}{C_1 C_2 R_1 R_2}}$$
(30)

4. For BSF, $V_3 = 0$ and $V_2 = V_1 = V_{in}$;

$$\frac{V_o}{V_{in}} = \frac{s^2 + \frac{K}{C_1 C_2 R_1 R_2}}{s^2 + s \frac{1}{R_3 C_2} + \frac{K}{C_1 C_2 R_1 R_2}}$$
(31)

5. For APF, $V_2 = V_1 = V_{in}$, $V_3 = -V_{in}$;

$$\frac{V_o}{V_{in}} = \frac{s^2 - s\frac{1}{R_3C_2} + \frac{K}{C_1C_2R_1R_2}}{s^2 + s\frac{1}{R_3C_2} + \frac{K}{C_1C_2R_1R_2}}$$
(32)

It is shown from the above equations that selecting the appropriate input voltages allows for the implementation of a five distinct second-order filter function. The filter parameters of the proposed circuit, namely, pole frequency (ω) and quality factor (Q) of this filter are given by:

$$Q = R_3 \sqrt{\frac{KC_2}{R_1 R_2 C_1}}$$
 (33)

$$\omega_o = \sqrt{\frac{K}{C_1 C_2 R_1 R_2}} \tag{34}$$

Here

$$K = \frac{R_f}{R_f + R_1} \tag{35}$$

3.2 SIMULATION AND EXPERIMENTAL RESULTS OF PROPOSED CIRCUIT:

To validate the functionality of the proposed filter, a comprehensive simulation study was conducted using PSPICE, leveraging the TSMC $0.18\mu m$ CMOS technology node.

In the simulation setup, the power supply voltages were set to V_{CC} =+5V, and V_{EE} =-5V. The universal filter was engineered to realize five fundamental filter responses: LPF, HPF, BPF, BSF and APF. The design targeted a center frequency of 1.4 MHz with a quality factor (Q) of 1. The selected component values were:

•
$$C_1 = C_2 = 150 pF$$
 (36)

$$\bullet \quad R=R_1=R_2=R_3=5K\Omega \tag{37}$$

•
$$R_f = 0.05 K\Omega$$
 (38)

Simulation results, depicted in Figures 3.2 through 3.6, illustrate the magnitude responses of the respective filters. The observed center frequency was approximately 1.31 MHz, slightly deviating from the theoretical value, likely due to inherent parasitic elements and non-idealities in the CMOS implementation.

To assess the robustness of the filter against component variations, a Monte Carlo simulation with 100 iterations was performed, considering a $\pm 10\%$ tolerance in capacitor values. The resulting frequency distributions, shown in Figures 3.7 to 3.9, demonstrate the filter's resilience and stability under manufacturing variances.

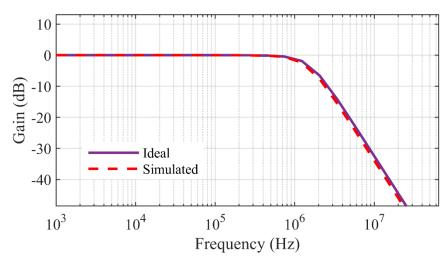


Figure 3.3: Theoretical and Simulated response of LPF

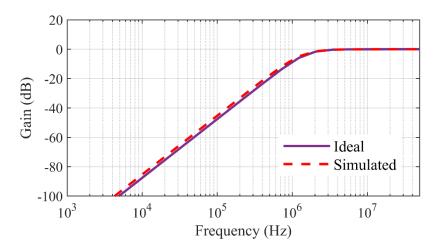


Figure 3.2: Theoretical and Simulated response of HPF

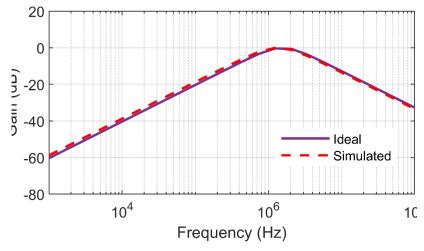


Figure 3.4: Theoretical and Simulated response of BPF

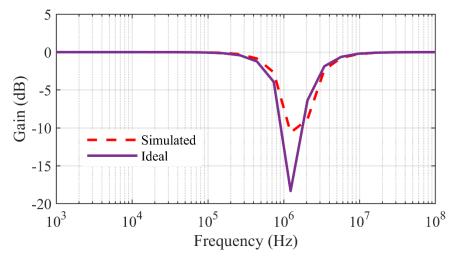


Figure 3.5: Theoretical and Simulated response of BRF

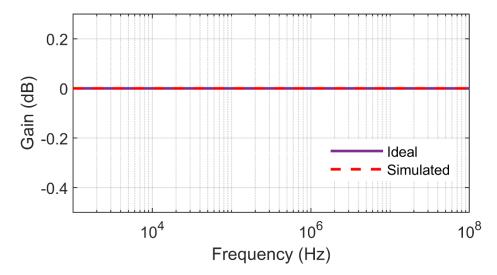


Figure 3.6: Theoretical and Simulated response of APF

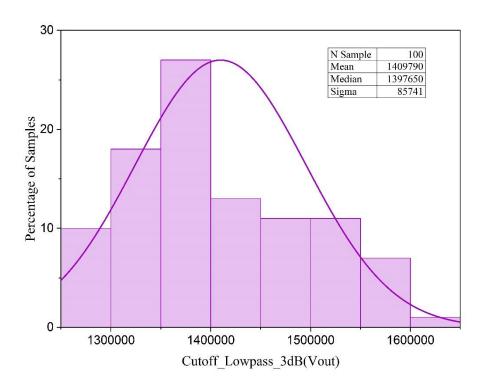


Figure 3.7: Monte Carlo analysis of LPF

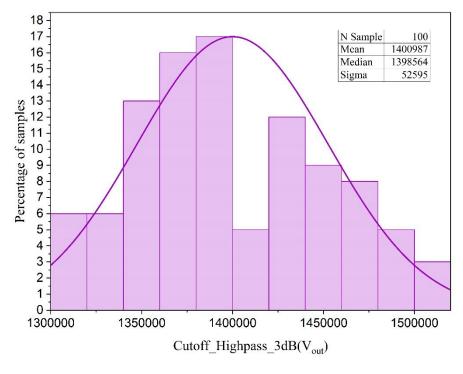


Figure 3.8: Monte Carlo analysis of HPF

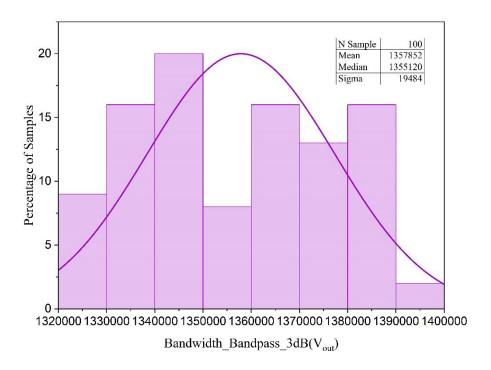


Figure 3.9: Monte Carlo analysis of BPF

CHAPTER 4

4. FUTURE SCOPE AND CONCLUSION

The **Second-Generation Voltage Conveyor (VCII)** has proven to be a groundbreaking component in the field of analog signal processing. Its ability to efficiently process both voltage and current-mode signals while addressing traditional challenges like power consumption, circuit complexity, and scalability makes it a preferred choice for modern applications. Through the course of this study, the capabilities of VCII have been extensively explored in the context of inverse filtering, multifunction filters, and transimpedance-mode filters.

The study analyzed pivotal VCII based filter designs that serve as benchmarks for innovative applications:

- 1. A reconfigurable multifunction filter showcasing the adaptability of VCII in performing LPF, HPF, and APF operations within a compact, power-efficient setup. The adjustable pass-band gain and stability across a wide operating range were notable achievements.
- 2. A transimpedance-mode all-pass filter using a single grounded capacitor and three VCIIs, which demonstrated exceptional phase-shifting capabilities with low harmonic distortion, making it particularly suited for portable electronics and sensor-based applications.
- 3. A current-mode third-order low-pass filter using a single grounded capacitor and seven VCIIs, designed with bulk-driven MOS transistors operating at 0.5 V, which demonstrated low power consumption, high linearity, and low harmonic distortion—making it highly suitable for biosensor and low-voltage portable biomedical applications.

These advancements collectively highlight the strength of VCII in enabling simpler, more efficient circuits while addressing modern-day challenges such as size constraints, power limitations, and integration feasibility.

4.1 FUTURE SCOPE OF THE PRESENTED WORK:

Despite its numerous advantages, there remains significant potential for further research and development in VCII based designs. Future efforts can focus on addressing specific limitations and expanding the application domains of VCII, as outlined below:

- 1. **Integration in Complex Systems**: While the reviewed designs are predominantly low-order filters, higher-order configurations could be explored. Such advancements would enable VCII-based circuits to cater to more sophisticated systems requiring higher selectivity, sharper transitions, or multi-band filtering capabilities.
- 2. **Digitally Controlled and Adaptive Circuits**: With the rise of smart systems and IoT devices, integrating digital control into VCII-based circuits could unlock new possibilities. This would allow real-time adaptability, enabling the circuits to dynamically adjust parameters like gain, frequency response, and phase shift based on system requirements.
- 3. Advanced Packaging and Miniaturization: As VCII finds applications in portable devices and biomedical systems, further research into reducing its size and enhancing its power efficiency is crucial. Efforts to integrate VCII into System-on-Chip (SoC) designs would be particularly beneficial, enabling more compact and energy-efficient solutions.
- 4. **Frequency Range Expansion**: While current designs cater to a moderate frequency range, extending the frequency capabilities of VCII-based circuits would make them suitable for high-frequency applications, such as radar systems, high-speed communication networks, and RF systems.
- 5. **Temperature and Process Variability**: Although VCII-based designs are generally robust, further work could focus on enhancing their performance under extreme environmental conditions. This would involve optimizing their sensitivity to temperature and process variations, making them more reliable for aerospace and automotive applications.
- 6. **Exploration in Mixed-Signal Systems**: VCII's capability to handle both voltage and current modes positions it uniquely for applications in mixed-signal environments. Integrating VCII with digital-to-analog and analog-to-digital converters could enhance system performance in areas like sensor fusion and real-time signal processing.
- 7. **Emerging Applications**: With the ongoing advancements in neural interfaces, quantum computing, and nanotechnology, VCII-based designs could find novel applications. Their adaptability and efficiency make them ideal candidates for the demanding requirements of these fields.

4.2 CONCLUSION:

In conclusion, the role of VCII in revolutionizing analog circuit design cannot be overstated. Its contributions to simplifying circuit architecture, enhancing efficiency, and enabling multifunctionality mark it as an indispensable tool for engineers and researchers. By addressing the outlined future directions, the full potential of VCII can be realized, paving the way for groundbreaking innovations in signal processing, communications, biomedical instrumentation, and beyond.

This study not only demonstrates the current capabilities of VCII but also sets the stage for its future exploration, ensuring its relevance in the ever-evolving landscape of electronics and system design.

REFERENCES

- [1.] Safari L., Barile G., Stornelli V., & Ferri G., "An overview on the second generation voltage conveyor: Features, design and applications", IEEE Transactions on Circuits and Systems II: Express Briefs, Vol. 66(4), pp. 547-551. (2018).
- [2.] Jaikla W., Sangyaem S., Supavarasuwat P., Khateb F., Minaei S., Kulej T., & Suwanjan P., "Reconfigurable voltage-mode first-order multifunction filter employing second-generation voltage conveyor (VCII) with complete standard functions and electronically controllable modification", IEEE Access, 11, pp. 56152-56169, (2023).
- [3.] Theppota B., Jaikla W., Saksiri W., & Siripruchyanun M., "VCII-Based Transimpedance-Mode All Pass Filter using Single Grounded Capacitor", In 2024 21st International Conference on Electrical Engineering/Electronics, Computer, Telecommunications and Information Technology (ECTI-CON) (pp. 1-6). IEEE, (2024, May).
- [4.] M. Kumngern, F. Khateb and T. Kulej, "0.5 V Current-Mode Low-Pass Filter Based on Voltage Second Generation Current Conveyor for Bio-Sensor Applications," in IEEE Access, vol. 10, pp. 12201-12207
- [5.] S. Kilinc and U Çam, "Transimpedance type fully integrated biquadratic filters using operational transresistance amplifiers," Analog Integrated Circuits and Signal Processing, vol. 47, pp. 193–198, 2006.
- [6.] I. Filanovsky, "Current conveyor, voltage conveyor, gyrator," In Proceedings of the 44th IEEE 2001 Midwest Symposium on Circuits and Systems. MWSCAS 2001, Dayton, OH, USA, 14–17 August 2001; pp. 314–317.
- [7.] A. Yesil and S. Minaei, "New simple transistor realizations of second-generation voltage conveyor," Int. J. Circuit Theory Appl., vol. 48, no. 11, pp. 2023–2038, Nov. 2020.
- [8.] G. Barile, G. Ferri, L. Safari, and V. Stornelli, "A new high drive class-AB FVF-based second generation voltage conveyor," IEEE Trans. Circuits Syst. II, Exp. Briefs, vol. 67, no. 3, pp. 405–409, Mar. 2020.

- [9.] T. Dostál and J. Pošpíil, "Hybrid models of 3-port immittance convertors and current and voltage conveyors," Electron. Lett., vol. 18, no. 20, pp. 887–888, 1982.
- [10.] J. G. Webster, Medical Instrumentation: Application and Design, 4th ed. Hoboken, NJ, USA: Wiley, 2010, pp. 11–12, (2010).
- [11.] S. Prakobnoppakao, B. Chipipop, W. Surakampontorn, and K. Watanabe, "Design of a Current-Mode CCII-Based Bandpass Filter From Immittance Function Simulator Using Commercial Available CCII(AD844)," in Proceedings of the ITC- CSCC (Phuket, Thailand: Kaewchai Chancharoen, 2002), pp. 743–746.
- [12.] H. Y. Wang and C. T. Lee, "Systematic Synthesis of RL and CD Immittances Using Single CCIII," International Journal of Electronics 87, no. 3 (2000): 293–301.
- [13.] D. Ozenli, E. Alaybeyoglu, and H. Kuntman, "A Tunable Lossy Grounded Capacitance Multiplier Circuit Based on VDTA for the Low Frequency Operations," Analog Integrated Circuits and Signal Processing 113, no. 2 (2022): pp.163–170.
- [14.] A. Abaci and E. Yuce, "Single DDCC Based New Immittance Function Simulators Employing Only Grounded Passive Elements and Their Applications," Microelectronics Journal 83 (2019): pp. 94–103.
- [15.] W. Tangsrirat and W. Surakampontorn, "Tunable Active Grounded Lossless and Lossy Inductance Simulators With Single Grounded Capacitor Using VDBAs," Scientia Iranica 29, no. 2 (2022), pp. 739–748.
- [16.] P. Mongkolwai, P. Moonmuang, T. Pukkalanun, and W. Tangsrirat, "Grounded Parallel RL, RC, and LC Immittance Simulators Using VDBAs and Passive Elements," in 2022 International Electrical Engineering Congress (iEECON) (Khon Kaen, Thailand: IEEE, 2022), 1–4.
- [17.] P. Moonmuang and W. Tangsrirat, "Electronically Tunable Resistorless Capacitance Multiplier Employing VDBAs," in Proceedings of the 3rd International Conference on Communication and Information Processing (Tokyo, Japan: Association for Computing Machinery, 2017), pp. 512–515.

- [18.] R. Bhagat, D. R. Bhaskar, A. Raj, and P. Kumar, "Single CDBA-Based Grounded Parallel Lossy Inductor Simulator Circuits," AEU International Journal of Electronics and Communications 168 (2023).
- [19.] A. Lahiri and A. Gupta, "Realizations of Grounded Negative Capacitance Using CFOAs," Circuits, Systems, and Signal Processing 30, no. 1 (2011) pp. 143–155.
- [20.] R. Verma, N. Pandey, and R. Pandey, "Novel CFOA Based Capacitance Multiplier and Its Application," AEU—International Journal of Electronics and Communications 107 (2019): 192–198.
- [21.] G. Di Cataldo, G. Ferri, and S. Pennisi, "Active Capacitance Multipliers Using Current Conveyors," in IEEE International Symposium on Circuits and Systems (ISCAS) (Monterey, CA: IEEE, 1998).
- [22.] D. V. Kamath, "Novel DO- OTA Based Current-Mode Grounded Capacitor Multiplier," in 2nd International Conference on Inventive Systems and Control (ICISC) (Coimbatore, India: IEEE, 2018).
- [23.] M. A. Al-Absi and M. T. Abuelma'atti, "A Novel Tunable Grounded Positive and Negative Impedance Multiplier," IEEE Transactions on Circuits and Systems II: Express Briefs 66, no. 6 (2018): 924–927.
- [24.] B. Aggarwal and A. Mittal, "Design of Temperature and Input Voltage Insensitive VDTA and Impedance Multiplier," Microelectronics Journal 85 (2019): 34–51.
- [25.] T. K. Paul, S. Roy, and R. R. Pal, "Lossy & Lossless Capacitance Multipliers: A Series of Realization Using VDTAs & Single Grounded Capacitor," Mapana Journal of Sciences 21, no. 3 (2022), pp. 1–34.
- [26.] D. R. Bhaskar, G. Mann, and P. Kumar, "OTRA-Based Positive/Negative Grounded Capacitance Multiplier," Analog Integrated Circuits and Signal Processing 111, no. 3 (2022), pp. 469–481.



16% Overall Similarity

The combined total of all matches, including overlapping sources, for each database.

Filtered from the Report

- Bibliography
- Quoted Text
- Cited Text
- Small Matches (less than 8 words)

Match Groups

118Not Cited or Quoted 16%

Matches with neither in-text citation nor quotation marks



99 0 Missing Quotations 0%

Matches that are still very similar to source material



0 Missing Citation 0%

Matches that have quotation marks, but no in-text citation



0 Cited and Quoted 0%

Matches with in-text citation present, but no quotation marks

Top Sources

Internet sources

Publications

Submitted works (Student Papers)

Integrity Flags

0 Integrity Flags for Review

No suspicious text manipulations found.

Our system's algorithms look deeply at a document for any inconsistencies that $% \left(1\right) =\left(1\right) \left(1\right) \left($ would set it apart from a normal submission. If we notice something strange, we flag it for you to review.

A Flag is not necessarily an indicator of a problem. However, we'd recommend you focus your attention there for further review.

