

Research Article

A Novel High-Performance Patch Radiator

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A novel two-layer highly efficient directive E-shaped patch radiator is described. By modifying the geometry of a rectangular patch and by introducing two slits, the size of the original rectangular patch is reduced. Further reduction in the size is achieved by stacking E-shaped patches. Both gain and efficiency of this modified antenna is increased by 16%. It is also observed that by introducing EBG structure, the bandwidth of the antenna is increased by 10.5% approximately.

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1. Introduction

Microstrip patch antennas (MSAs) [1–3] have attractive features such as low weight, low profile planar configurations, ease of fabrication, and integration with RF devices which are being used for many applications due to key advantages over conventional antennas. However, there are several applications where even the physically small size of a conventional MSA is still too large. One such application is handset of mobile communication. Although the above said features are compatible with the mobile communication technology, the operational frequency is too low. Also the practical limitations of these MSAs are low gain and low efficiency. Thus, several techniques have been proposed to alleviate these problems, such that effectively the size of the radiator is reduced. One such method is to use a thicker substrate with low-permittivity constant, which is not acceptable for mobile communications. Other proposed methods include using high dielectric constant material [4], multiple resonances, and modifying the geometry of patch antennas [5–9]. However, using high dielectric constant material, the efficiency was poor and the bandwidth is narrow due to surface wave excitation. Modifying the geometry of the patch [10, 11] instead gives better results. Size of the microstrip patch antenna can be reduced by using high-permittivity substrate or shorting posts at the cost of bandwidth, gain, and efficiency.

It is difficult to achieve high gain and high efficiency from simple rectangular MSA. To fulfill these requirements,

the geometry of a rectangular patch antenna has to be modified as discussed below. A pair of slits is inserted into an appropriate radiating edge of the rectangular patch antenna. These slots reduce the size of the original rectangular patch because the length of the current path around the slots is increased [12]. To achieve this, a rectangular microstrip antenna of size 105 mm \times 44 mm is considered and is fed with a microstrip line at the center of the long edge of the antenna. Two parallel narrow slots of length and width 43.25 mm \times 1 mm are inserted symmetrically into the radiating edge of a normal rectangular patch resembling the English letter E. Also to achieve bandwidth enhancement [13, 14], a patch is etched on one side of a dielectric substrate micro machined [15] with a rectangular lattice of holes on the ground plane. These lattices of holes act as electromagnetic bandgap (EBG) structure.

Two designs of microstrip patch antennas are studied. First, antenna design has an E-shaped stacked microstrip antenna loaded with three shorting posts. Second, design has an E-shaped antenna with electromagnetic bandgap (EBG) structure etched on the other side of the substrate.

2. Proposed New Designs

The designs proposed incorporate novel methods for additional, independently radiating resonances within the structure. Also they are fabricated on thick substrate with low dielectric constants which have enough bandwidth for 3G applications.

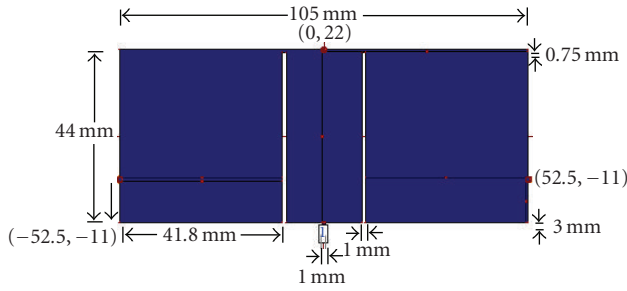


FIGURE 1: Top view of LS ESP antenna.

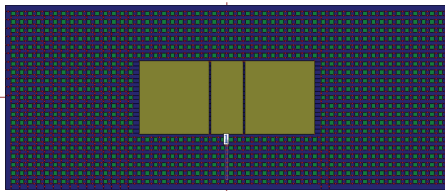


FIGURE 2: Top view of EBG ESP.

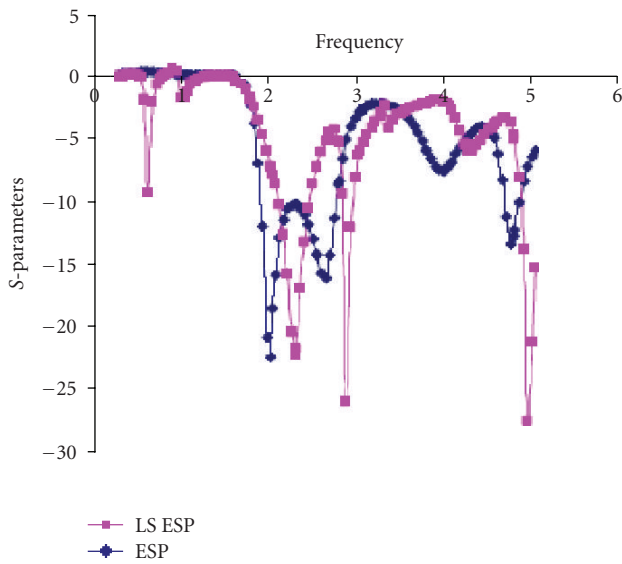
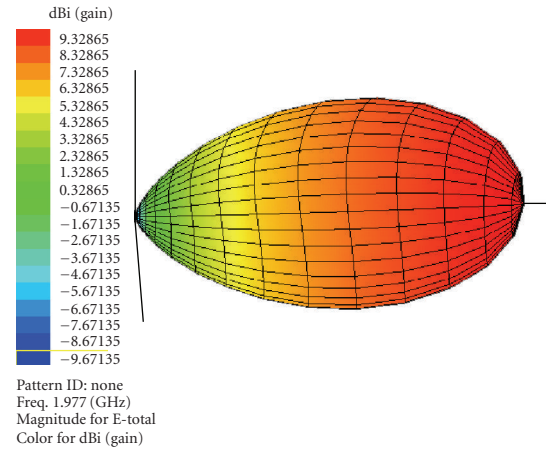


FIGURE 3: Simulated S-parameters of the ESP and LS ESP.

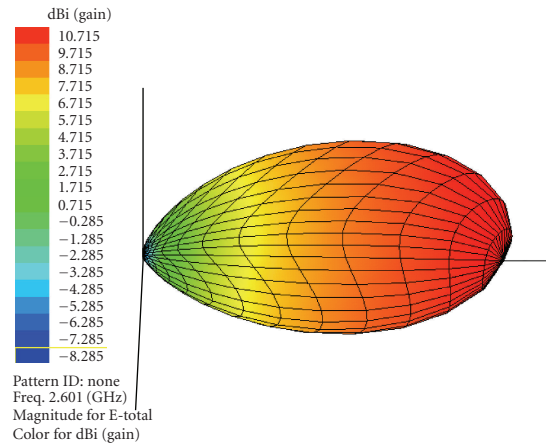
2.1. Antenna Design-I

A conventional rectangular microstrip antenna's resonant frequency can be reduced by placing a shorting post at the center of the edges. However, this shorted rectangular patch antenna has low gain and efficiency. To improve these parameters, a stacked E-shaped patch is used in the first design.

The geometry of loaded stacked E-shaped patch antenna (LSESP) consists of two identical E-shaped patches (ESPs). The design of the antenna consists of two air dielectrics of 18 mm height each. The first ESP is placed at 18 mm above the ground plane, and the second ESP is stacked on the first patch at 36 mm from the ground plane. Two shorting posts are connected from the ground to the lower patch at the



(a)



(b)

FIGURE 4: (a)-(b) Radiation patterns of the ESP at 1.977 GHz and 2.601 GHz.

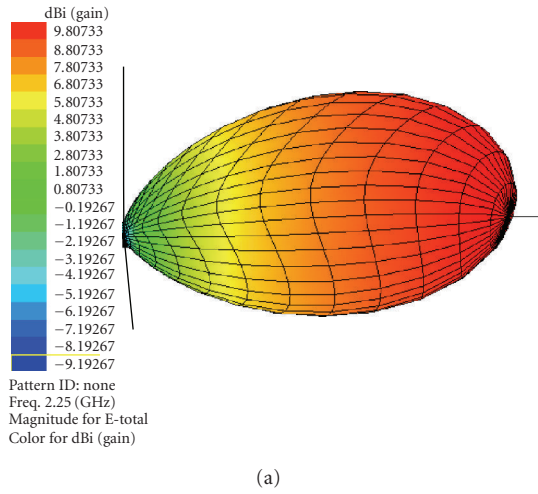
points (52.5, -11) and (-52.5, -11), and the third post is connected from the lower to the upper patch at the point (0,22) as shown in Figure 1.

The excitation for the antenna is given by a microstrip line feed to the lower patch.

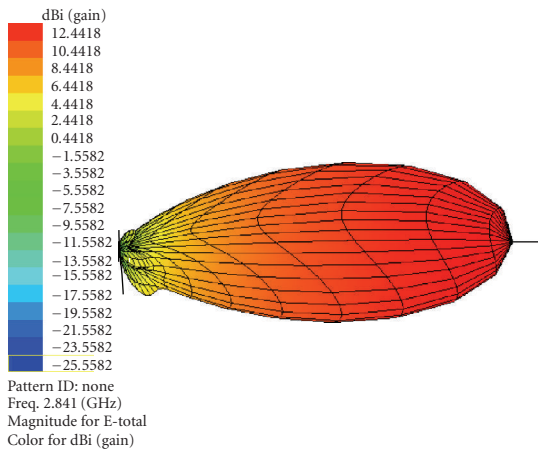
2.2. Antenna Design-II

In the second design, the antenna consists of a single dielectric layer on which E-shaped patch is etched on one side.

On the other side, a 265 mm × 110 mm rectangular metallic ground plane is constructed. Slots of 2.5 mm × 2.5 mm square holes spaced 5 mm apart forming a 52 × 21 matrix have been made on the ground plane. There are no slots in the area just below the patch. This rectangular ground plane acts as EBG [16] ground plane. The excitation for the antenna is given by a microstrip line feed to the patch as shown in Figure 2.



(a)



(b)

FIGURE 5: (a)-(b) Radiation patterns of the LS ESP at 2.25 GHz and 2.841 GHz.

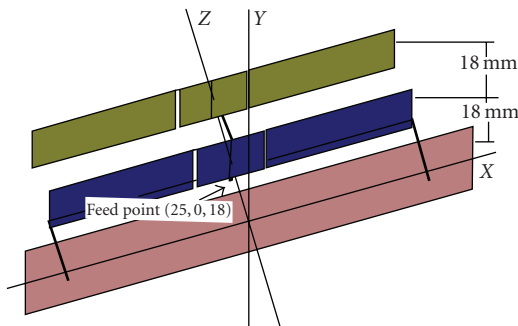


FIGURE 6: 3D view of the E-shaped patch.

3. Results and Discussions

To simulate the structures, IE3D software [17] is used, which is based on method of moments. The first design is used to provide high gain and efficiency, whereas the second design

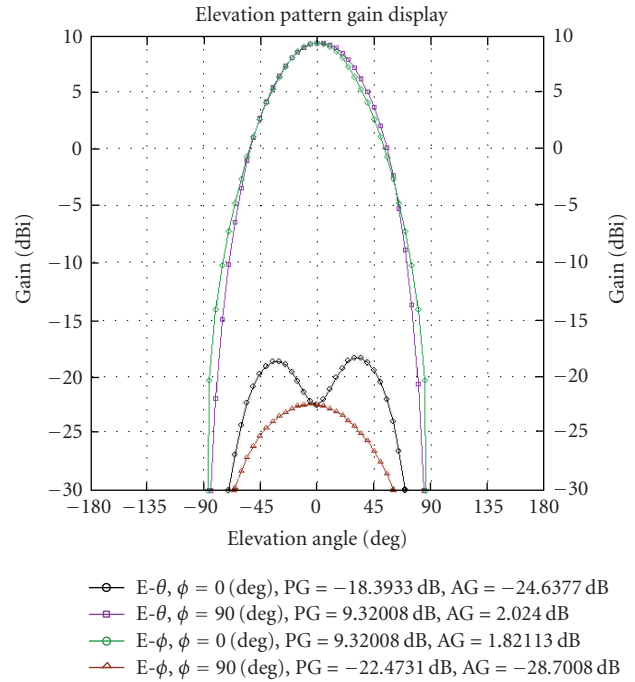


FIGURE 7: 2D pattern of the ESP at 1.977 GHz.

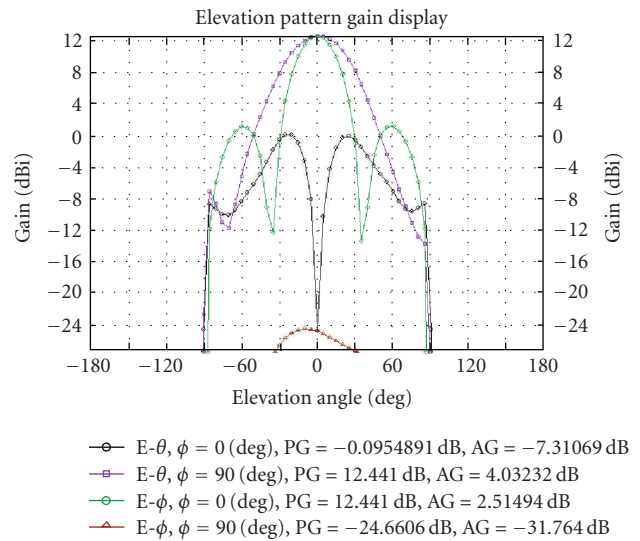


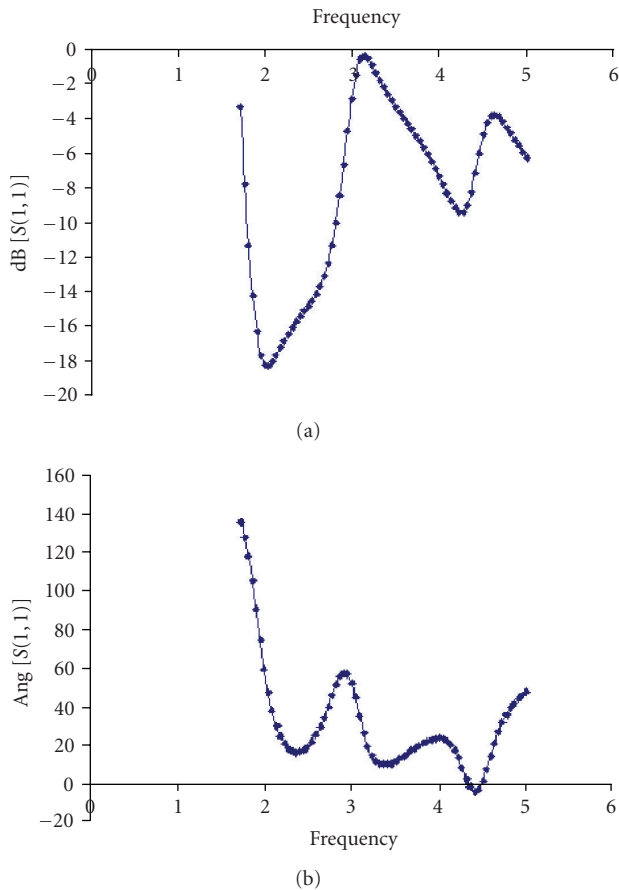
FIGURE 8: 2D pattern of the LS ESP at 2.84 GHz.

provides larger bandwidth as discussed below. The various parameters of ESP and LS ESP have been tabulated in Table 1.

As seen from Table 1, both resonant frequencies of ESP antenna have been increased in case of LS ESP as shown in Figure 3. Radiation and antenna efficiency are increased by 16% in case of first resonant frequency, and in case of second resonant frequency both of them are almost the same as shown in Figures 4 and 5. Also, the gain and directivity of the LS ESP have been increased by 16% in second resonant frequency. Further, the 2D patterns of the ESP and LS ESP are shown in Figures 7 and 8.

TABLE 1: Comparison of ESP and LS ESP antennas.

Antenna design	Resonant frequency (GHz)	VSWR	Radiation efficiency (%)	Antenna efficiency (%)	Gain (dBi)	Directivity (dBi)	3 dB beam width
ESP	1.98	1.16	83.21	82.78	9.34	10.15	60.27, 65
LS ESP	2.25	1.16	99.83	99.30	9.81	9.84	38.04, 69.91
ESP	2.60	1.36	100	97.7	10.72	10.82	35.69, 64.57
LS ESP	2.84	1.10	100	97.11	12.44	12.57	34.16, 48.54

FIGURE 9: Simulated $\text{dB}[S(1,1)]$ and $\text{Ang}[S(1,1)]$ versus frequency of ESP EBG.

The impedance bandwidth calculated by the EBG ESP antenna is 10.5% more than the bandwidth provided by the simple E-shaped patch radiator. The S-parameters of the EBG ESP antenna are as shown in Figure 9.

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The demand for new generation wireless networks has spurred a vibrant flurry of research on cooperative communications during the last few years. Nevertheless, many aspects of cooperative communications are open problems. Furthermore, most of the cooperative systems proposed so far are based on ideal assumptions, such as unfeasible synchronization constraints between the relay nodes or the availability of perfect channel state information at the resource allocation unit. There is a need for research on practical ways of realizing cooperative schemes based on realistic assumptions.

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